



Prediction-based resource allocation in OFDMA

J. F. Schmidt, J. E. Cousseau, R. Wichman S. Werner

Universidad Nacional del Sur, Argentina

Aalto University School of Electrical Engineering, Finland



Introduction

Motivation

- OFDMA is a popular physical layer technique (WiMAX, LTE)
- Resource allocation (RA) by physical layer scheduling
 - Increases system throughput
 - Fairness issue
 - Outdated channel state information in the transmitter (CSIT)
- Prediction-based RA has been shown to improve fairness
 - No consideration of prediction error characteristics so far
 - Prediction error increases with prediction horizon

Contributions

- Analysis of the prediction horizon
 - Length vs. accuracy
 - Characterization of the prediction error
 - Many samples of the prediction error in OFDMA
 - Histograms to approximate error statistics
 - No specific error model assumed
 - Applies to general long-range channel predictors
 - Realistic simulation parameters
 - LTE parameters
 - ITU Vehicular A channel
 - Practical low-complexity long-range channel predictor
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System model

- OFDMA downlink transmission
- N subchannels \rightarrow allocation in the resolution of 1 subcarrier
- K active mobile stations (MSs)
 - Independent fading channels, same statistics
- Time slot based resource allocation
 - M OFDM symbols = 1 time slot
 - Channel approximately constant for 1 time slot
- Transmitted power equally distributed among subchannels

Prediction-based resource allocation

$$R_k(s) = M \sum_{n \in \mathcal{I}_k} r_k(s, n) \quad \longrightarrow \quad \text{Rate achieved by user } k \text{ on time slot } s$$

$n \in \mathcal{I}_k$ \longrightarrow Subchannels allocated to MS k

Prediction-based PFS (P-PFS) checks W future time slots (TS) when assigning the next time slot such that

$$P^{(s+1)}(\bar{R}_k^W) = \arg \max_{\mathcal{P}} \sum_{k=1}^K \log(\bar{R}_k^W)$$

where

$$\bar{R}_k^W = \left(1 - \frac{1}{\tau}\right) \bar{R}_k(s) + \frac{1}{\tau} \sum_{w=1}^W \left(1 - \frac{1}{\tau}\right)^{W-w} \check{R}_k(s+w)$$

$\bar{R}_k(s)$ \longrightarrow Average rate
 $\check{R}_k(s+w)$ \longrightarrow Achievable rate

P-PFS relies on the predicted achievable rates

Error-aware prediction-based resource allocation

Error-aware prediction-based RA

Achievable rates are computed to meet the target BER as

Perfect CSIT	Imperfect CSIT
$P_e(s, n) \approx c_1 \exp \left\{ \frac{-c_2 \gamma H(s, n) ^2}{2^{\beta(s, n)} - 1} \right\}$	$\bar{P}_e(s, n) = E_{ H(s, n) }^{\hat{H}(s, n)} \{P_e(s, n)\}$
<p>Mean SNR</p> <p>Bits/symbol</p>	<p>Estimated/predicted CSIT</p>

W pdfs needed to predict W time slots

Closed form solutions exist for the estimated but not for the predicted CSIT

Error-aware prediction-based RA

Expressing $\bar{P}_e(s, n)$ for each w in terms of the prediction error e_w

$$\bar{P}_{e_w}(s+w, n) = \hat{P}_{e_w}(s+w, n) \cdot \rho_w(n) \rightarrow \text{Correction factor}$$

where

$$\rho_w(n) = \frac{1}{\sigma_H^2} \int_{-\infty}^{\infty} \exp \left\{ \frac{-c_2 \gamma \theta(s+w, n)}{2^{\beta(s+w, n)} - 1} \right\} f_{e_w}(e_w) de_w$$

$$\theta(s+w, n) = 2\sigma_H^2 \left| \hat{H}(s+w, n) \right|_{|H(s, n)} e_w + \sigma_H^4 e_w^2$$

with

Error-aware prediction-based RA

Error samples for different subcarriers can be used to construct a histogram to estimate $f_{e_w}(e_w)$

Defining the intervals $\Delta_i = \left\{ e_w : \epsilon_i - \frac{\omega}{2} < e_w \leq \epsilon_i + \frac{\omega}{2} \right\}$

Interval width

Histogram thresholds

Results in $\hat{f}_{\epsilon_i}(e_w) = \frac{Q_i}{Q_w}$ for $e_w \in \Delta_i$

Number of samples within the i th interval

Total number of samples

such that the correction factor can be rewritten as

$$\hat{\rho}_w(n) = \frac{1}{\sigma_H^2} \sum_{-\epsilon_Q}^{\epsilon_Q} \exp \left\{ \frac{-c_2 \gamma \tilde{\theta}(s+w, n)}{2^{\beta(s+w, n)} - 1} \right\} \hat{f}_{\epsilon_i}(\epsilon_i)$$

$$\bar{P}_{e_w}(s+w, n) = \hat{P}_{e_w}(s+w, n) \cdot \hat{\rho}_w(n)$$

Channel estimator/predictor

Channel estimator/predictor

BEM Estimators	Kalman Estim/pred
Good fit to practical channels (Robust to Doppler shape)	Good fit to practical channels (Robust to Doppler shape)
Not suitable for prediction	Good prediction performance
Low computational cost	High computational cost

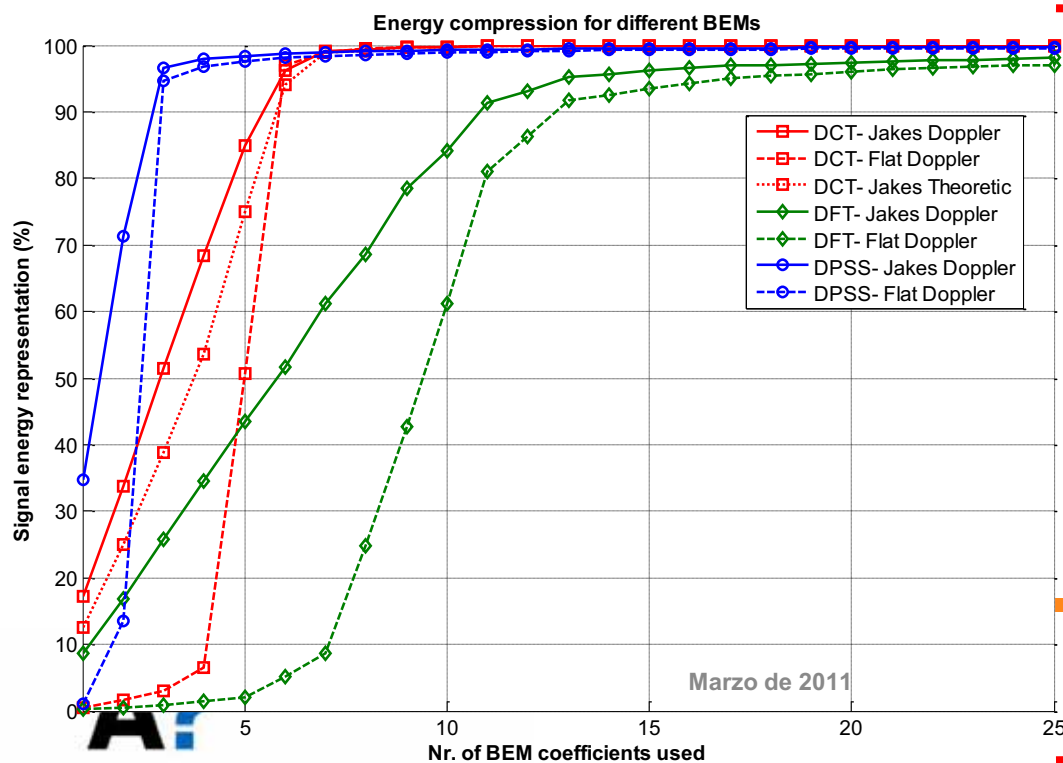
Recursive basis expansion model (BEM) estimator/predictor

- Good fit to practical channels (Robust to Doppler shape)
- Good prediction performance
- Low computational cost

Recursive Basis Expansion Model

- Evolution of the temporal channel can be expressed using BEM $\hat{\mathbf{H}} = \bar{\mathbf{T}}^H \bar{\boldsymbol{\gamma}}$ as
- Different BEMs lead to different compressions

$$E \left\{ \sum_{m=0}^{M-1} |H(m)|^2 \right\} = E \left\{ \sum_{m=0}^{M-1} |\hat{H}(m)|^2 \right\} + \sigma_e^2$$



DCT → Good compression and simple basis functions

Recursive BEM estimator

- For DCT $\rightarrow [\bar{\mathbf{T}}]_{i,m} = A(m) \cos\left(\frac{\pi(i + \frac{1}{2})m}{M}\right)$

- $\bar{\mathbf{T}}$ can be represented with a filter bank as

$$\hat{H}_{(DCT)}(m) = H_{DCT}(q^{-1})H(m) = \sum_{i=0}^{G-1} H_{DCT_i}(q^{-1})H(m)$$

where

$$H_{DCT_i}(e^{j\omega}) = c_i \frac{(-1)^i - (-1)^i e^{-j\omega} - e^{-j\omega M} + e^{-j\omega(M+1)}}{1 + 2 \cos(i\pi/M) e^{-j\omega} + e^{-j2\omega}}$$

which can in turn be approximated by

$$H_F(e^{j\omega}) = \beta_0 \frac{0,5(1 - s_{20})(1 + e^{-j2\omega})}{1 - s_{20}e^{-j2\omega}} + \sum_{i=1}^{G-1} \beta_i \frac{0,5(1 - s_{2i})(1 - e^{-j2\omega})}{1 + (s_{2i} + 1)s_{1i}e^{-j\omega} + s_{2i}e^{-j2\omega}}$$

Recursive DCT BEM estimator

The filter bank can be inserted into a Kalman formulation having a steady-state solution as the basis set is time invariant

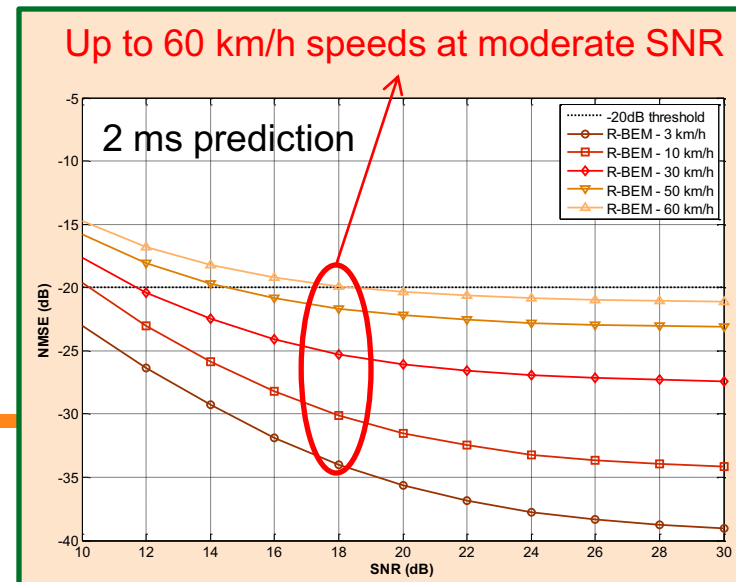
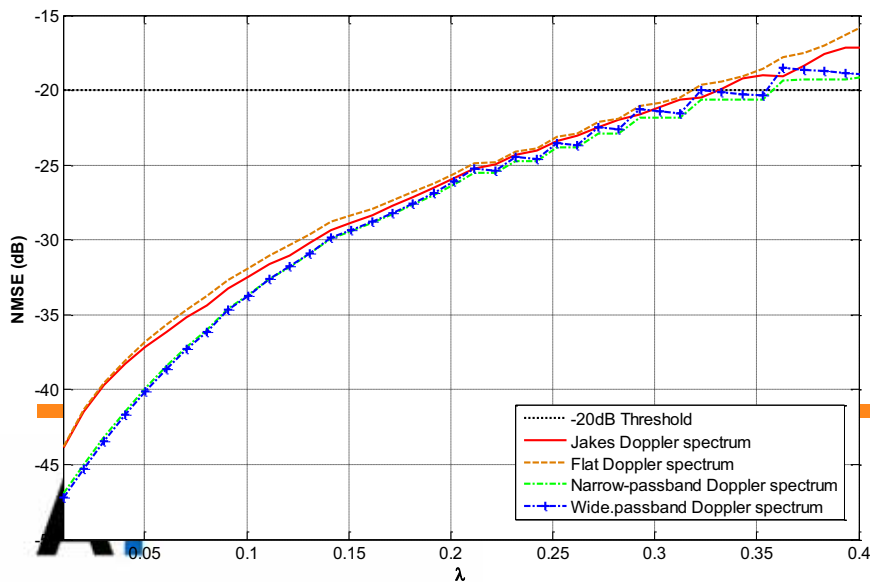
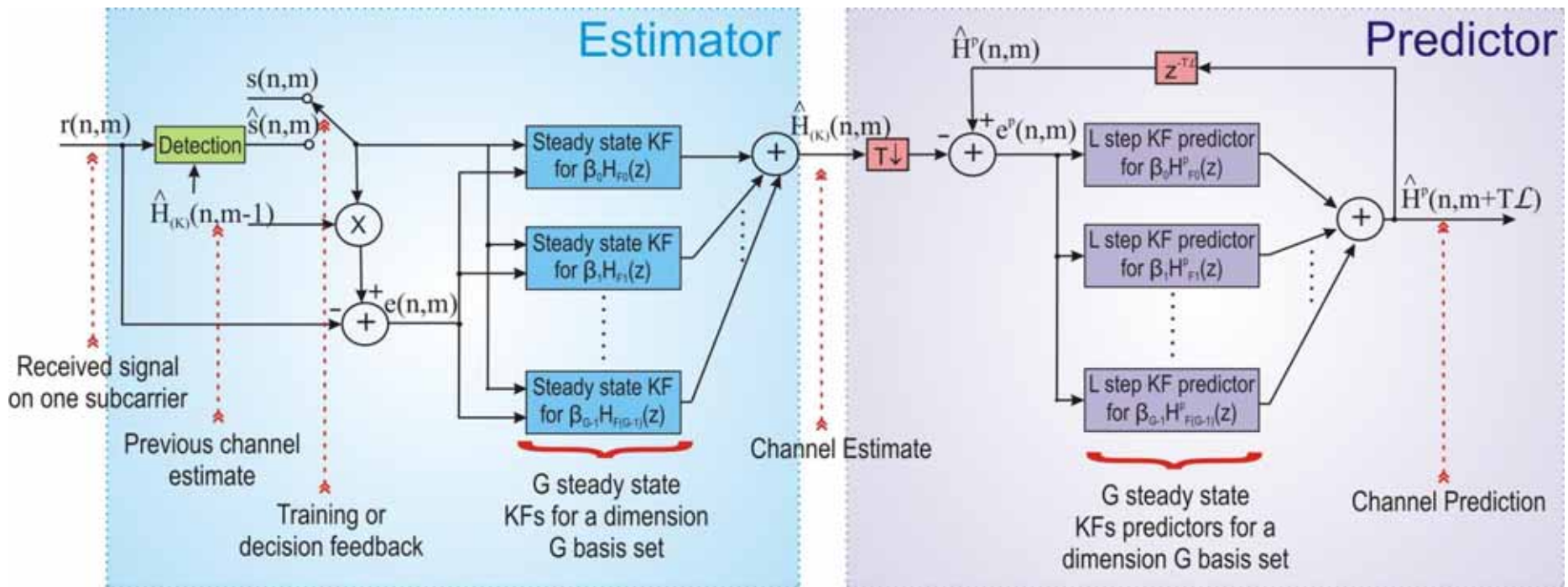
$$\left[\begin{array}{l} \mathbf{x}_i(m+1) = \mathbf{A}_i \mathbf{x}_i(m) + \mathbf{b}_i \bar{H}(m) \\ \hat{H}_{i(F)}(m) = \mathbf{c}_i \mathbf{x}_i(m) + d_i \bar{H}(m) \\ \hat{H}_{(F)}(m) = \sum_{i=0}^{G-1} \hat{H}_{i(F)}(m), \end{array} \right. \quad \left. \begin{array}{l} e(m) = r(m) - \hat{s}(m) \hat{H}_{(K)}(m-1) \\ \mathbf{x}_i(m) = \mathbf{A}_i \mathbf{x}_i(m-1) + \mathbf{k}_i \hat{s}^*(m) e(m) \\ \hat{H}_{i(K)}(m) = \mathbf{c}_i \mathbf{x}_i(m) \\ \hat{H}_{(K)}(m) = \sum_{i=0}^{G-1} \hat{H}_{i(K)}(m). \end{array} \right.$$

Recursive BEM long-range predictor

- Doppler bandwidth \ll OFDM bandwidth
- Temporal evolution \rightarrow highly oversampled

} Decimation in time and extrapolation of the Kalman filter

$$\begin{aligned} e^p(m+\ell) &= \hat{H}_{(K)}(m+\ell-T) - \hat{H}^p(m+\ell-T) \\ \mathbf{x}_{i\ell}^p(m+\ell) &= (\mathbf{A}_i^p)^{\mathcal{L}} \mathbf{x}_{i\ell}^p(m+\ell-T\mathcal{L}) + \mathbf{k}_i^p e^p(m+\ell) \\ \hat{H}_i^p(m+\ell) &= \mathbf{c}_i^p \mathbf{x}_{i\ell}^p(m+\ell) \\ \hat{H}^p(m+\ell+T(\mathcal{L}-1)) &= \sum_{i=0}^{G-1} \beta_i \hat{H}_i^p(m+\ell), \end{aligned}$$



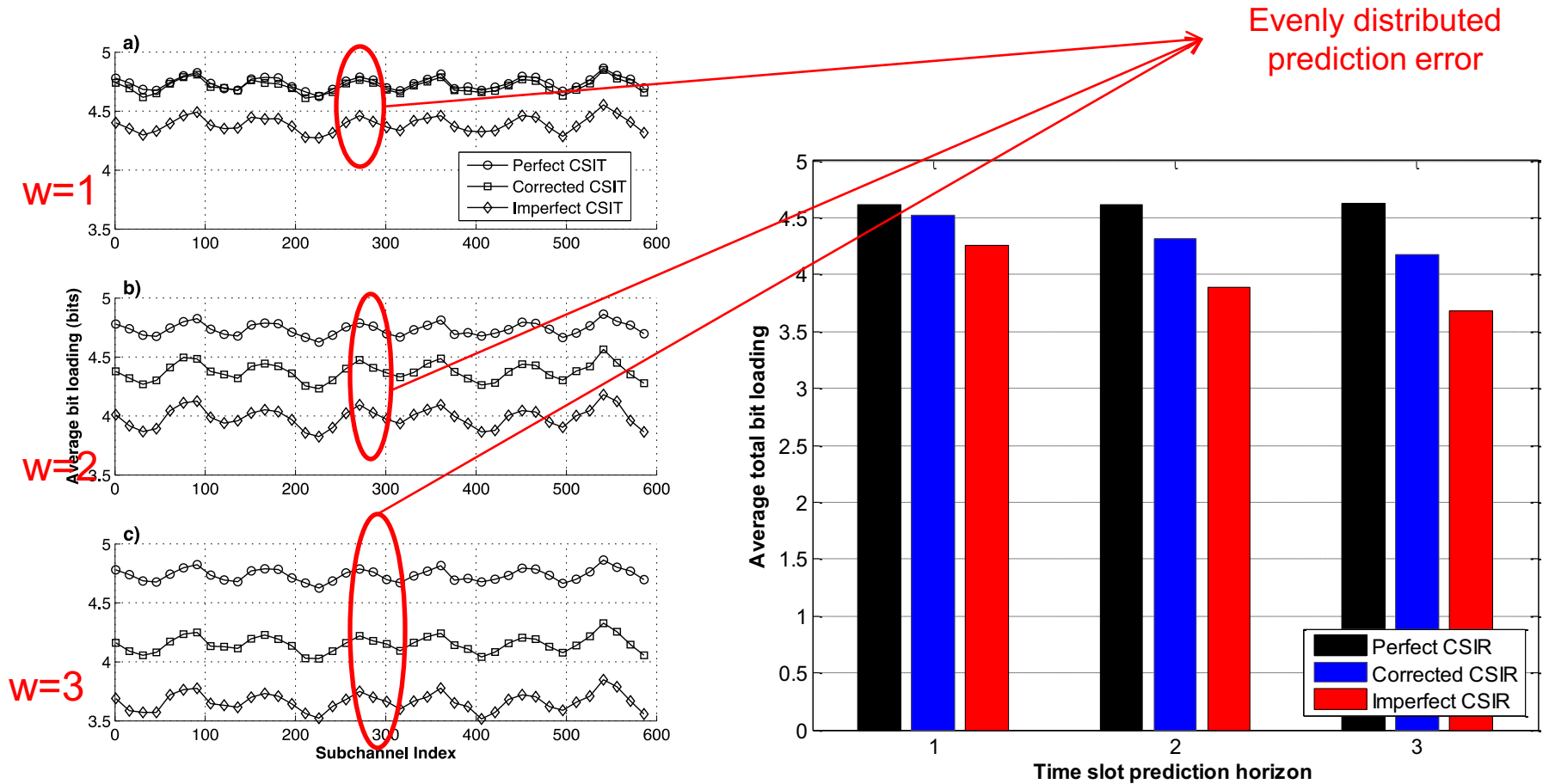
Numerical Results

Numerical results

System (LTE) parameters	Histograms parameters
<ul style="list-style-type: none">•Carrier frequency \rightarrow 2 GHz•Bandwidth \rightarrow 10 MHz•Subcarrier spacing \rightarrow 15 KHz•Cyclic prefix \rightarrow 5 μs•FFT size \rightarrow 1024•$N = 600$ subcarriers (90% bandwidth)•$M = 15$ OFDMA symbols = 1 ms•$W = 3$-time slot prediction•4, 16 and 64 QAM available•Target BER = $1 \cdot 10^{-3}$•ITU-Vehicular A channel 60 km/h•SNR \rightarrow 25 dB	<ul style="list-style-type: none">•$Q = 600$ (all subchannels used)•$\omega = 0.078$•Histogram limits ± 0.8

Numerical results

Single user, degradation of bit loading



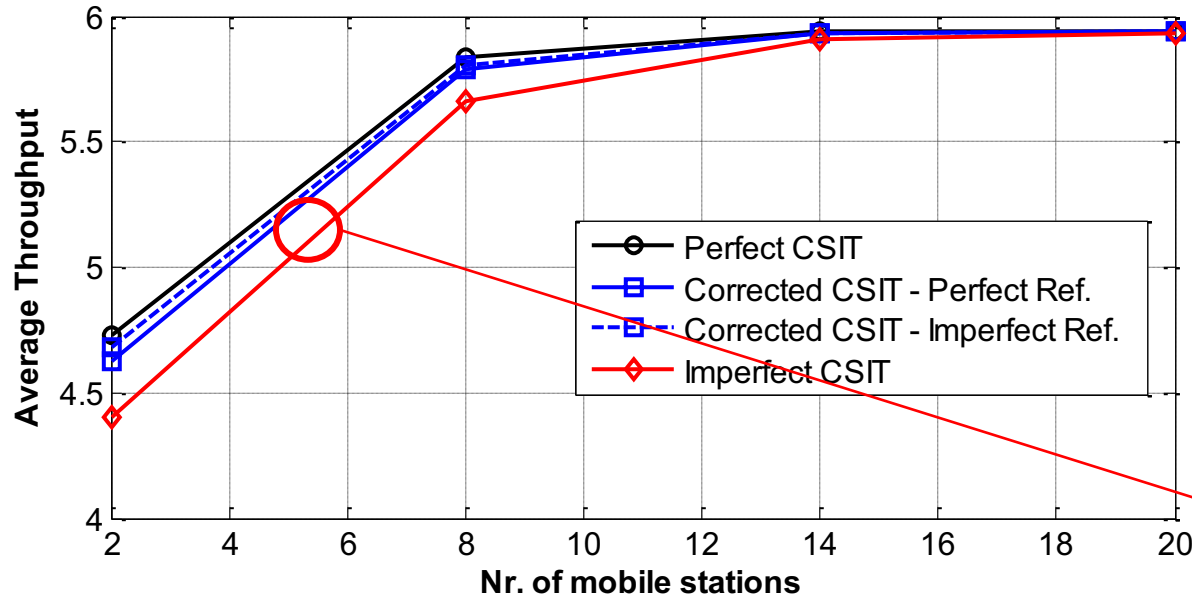
Evenly distributed prediction error

Numerical results

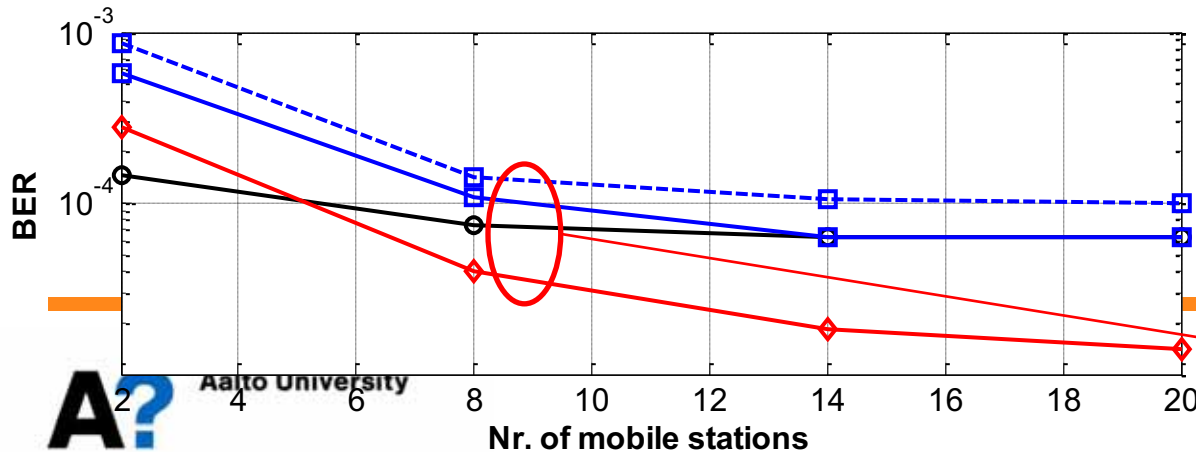
Impact to bit loading

Total system throughput

$$R(s) = \sum_{k=1}^K R_k(s)$$



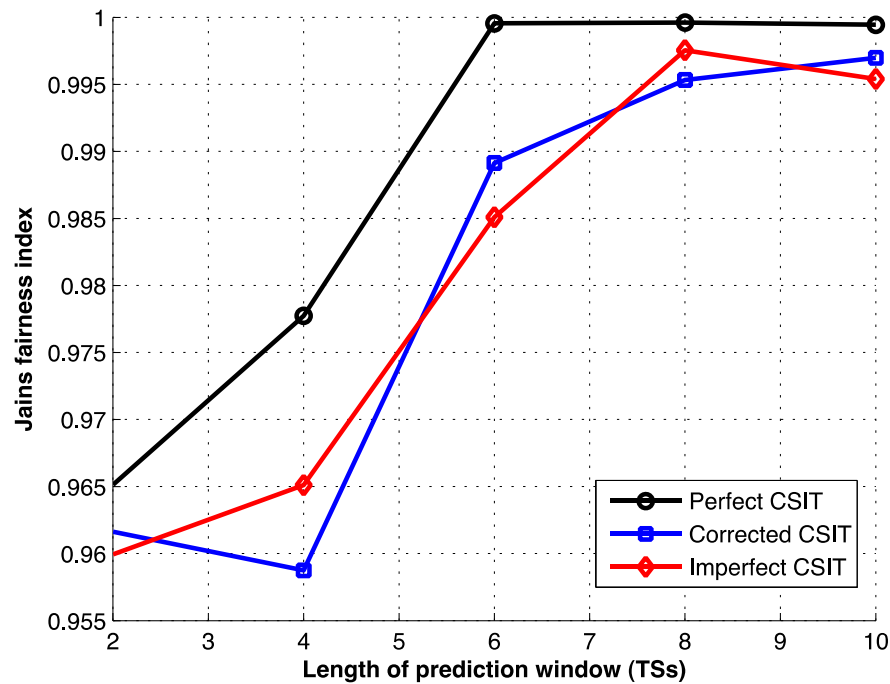
Reduced degradation of system throughput



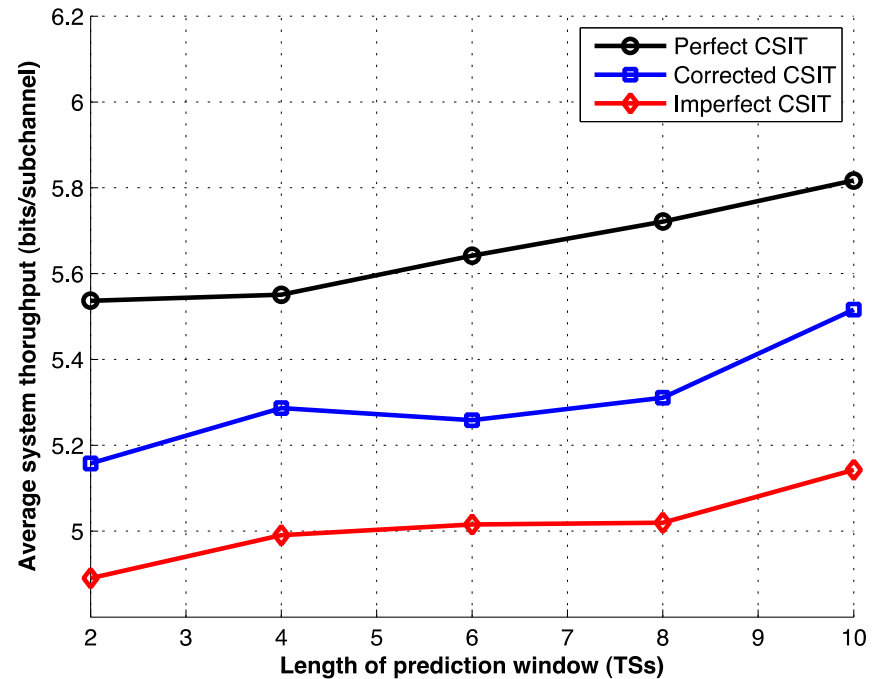
BER improvement

Numerical Results

Fairness



Throughput



$$\mathcal{J}(x_1, x_2, \dots, x_n) = \frac{(\sum_{i=1}^n x_i)^2}{n \cdot \sum_{i=1}^n x_i^2}$$

Conclusion

Conclusion

- A scheme to compensate the prediction error was proposed for the prediction-based resource allocation within mobile OFDMA.
- The scheme was evaluated under realistic system conditions.
- The scheme outperforms those that disregard prediction error.



Aalto University
School of Electrical
Engineering

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